

Systems Analysis and Control

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Lecture 13: Root Locus Continued

Overview

In this Lecture, you will learn:

Review

- Definition of Root Locus

Points on the Real Axis

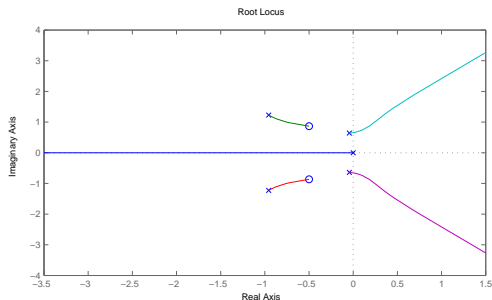
- Symmetry
- Drawing the Real Axis

What Happens at High Gain?

- The effect of Zeros
- Asymptotes

Root Locus

Review



Definition 1.

The **Root Locus** of $G(s)$ is the set of all poles of

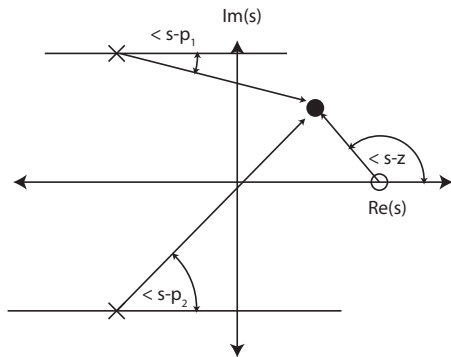
$$\frac{kG(s)}{1 + kG(s)}$$

as k ranges from 0 to ∞

Root Locus

Review

$$G(s) = \frac{(s - z_1) \cdots (s - z_m)}{(s - p_1) \cdots (s - p_n)}$$



For a point on the root locus:

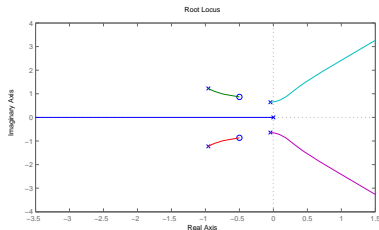
$$\angle G(s) = \sum_{i=1}^m \angle(s - z_i) - \sum_{i=1}^n \angle(s - p_i) = 180^\circ \pm n \cdot 360^\circ$$

Root Locus Demo 1

Wiley+ Root Locus Demo 1

Property of Symmetry

Constructing the Root Locus



Symmetry:

- Complex roots come in pairs: $a \pm bi$.
- Points on the root locus are mirrored above/below the real axis

We can divide points on root locus into

- Points on the real axis
- Symmetric Pairs off the real axis.

Points on the Real Axis

Phase Contribution from Zeros

What is the **PHASE** at a point ($s = a$) on the Real Axis?

We need:

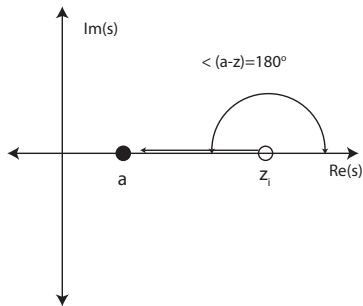
$$\sum_{i=1}^m \angle(a - z_i) - \sum_{i=1}^n \angle(a - p_i) = 180^\circ \pm n \cdot 360^\circ$$

Phase Contribution from Zeros (z_i):

Case 1: If z_i is on the Real Axis

- If $a > z_i$, then $\angle(a - z_i) = 0^\circ$
 - ▶ a is right of the zero.
- If $a < z_i$, then $\angle(a - z_i) = -180^\circ$
 - ▶ a is left of the zero.

Contribution is 0° or 180° !



Root Locus on the Real Axis

Phase Contribution of Complex Zeros

Case 2: If the z_i is Complex: Complex Roots come in pairs.

$$z_{1,2} = b \pm ci$$

- First Zero: Trigonometry
($a - z_1 = a - b + ci$)

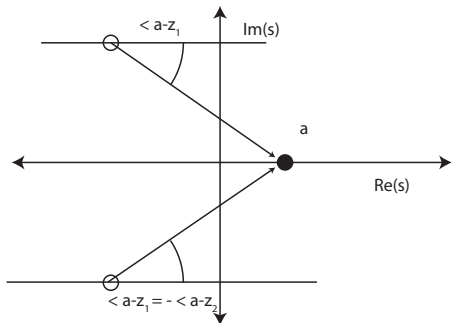
$$\angle(a - z_1) = \tan^{-1} \left(\frac{a - b}{c} \right)$$

- Second Zero:

$$\begin{aligned} \angle(a - z_2) &= \tan^{-1} \left(-\frac{a - b}{c} \right) \\ &= -\angle(a - z_1) \end{aligned}$$

Since $\angle(a - z_1) = -\angle(a - z_2)$, the total contribution to phase is 0° !

$$\angle(a - z_1) + \angle(a - z_2) = 0^\circ$$



Root Locus on the Real Axis

Phase Contribution from Real Poles

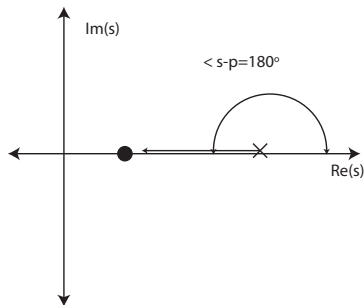
The Poles

$$\sum_{i=1}^m \angle(a - z_i) - \sum_{i=1}^n \angle(a - p_i) = -180^\circ$$

Similar to zeros

Case 1: If p_i is Real

- If a is right of p_i , then $\angle(a - p_i) = 0^\circ$
- If a is left of p_i , then $\angle(a - p_i) = -180^\circ$



Root Locus on the Real Axis

Phase contribution from Complex Poles

Case 2: If the p_i are Complex: Complex Roots some in pairs.

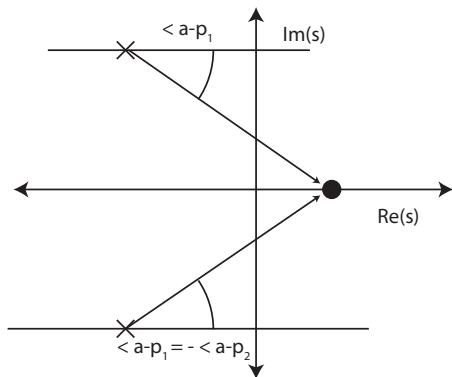
$$z_{1,2} = b \pm ci$$

- First Pole:

$$\angle(a - p_1) = \tan^{-1} \left(\frac{a - b}{c} \right)$$

- Second Pole:

$$\angle(a - p_2) = -\angle(a - p_1)$$



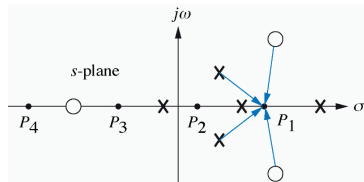
Again, $\angle(a - p_1) = -\angle(a - p_2)$ so **the Contribution is 0° !**

Root Locus on the Real Axis

ON-OFF

Summary: A point on the real axis: $s = a$

- Complex poles and zeros don't matter
- Real poles and zeros contribute 0° or 180°
 - ▶ 0° if the pole/zero is to the left of a
 - ▶ 180° if the pole/zero is to the right of a



The **PHASE** of $G(a)$ is

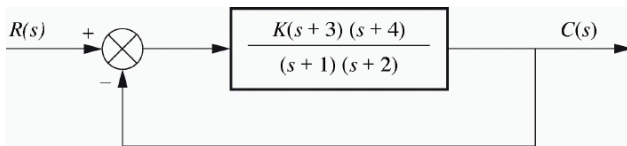
$$\angle G(a) = 180^\circ \cdot (\# \text{ of poles and zeros to the right of } a)$$

A Simple Rule:

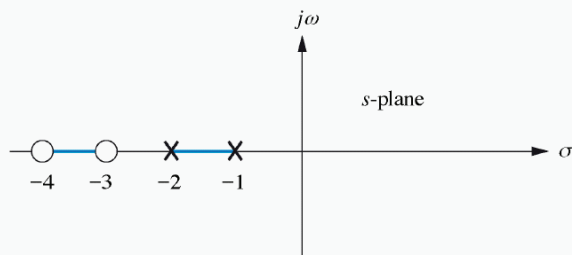
- If the # of poles and zeros to the right of a is **EVEN**.
 - ▶ We are **OFF** the root locus.
- If the # of poles and zeros to the right of a is **ODD**.
 - ▶ We are **ON** the root locus.

Root Locus on the Real Axis

Examples



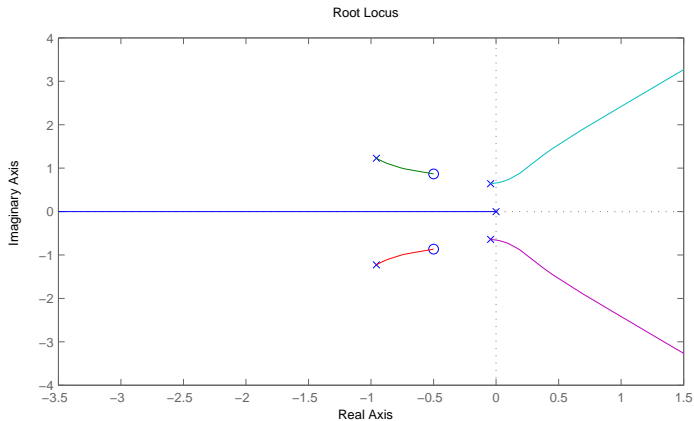
(a)



(b)

Root Locus on the Real Axis

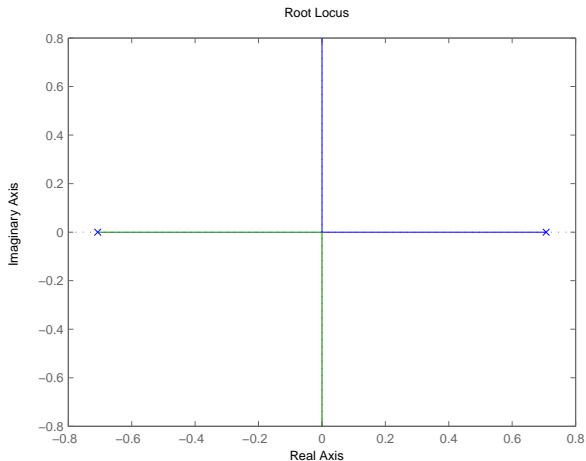
Examples



Root Locus on the Real Axis

Examples

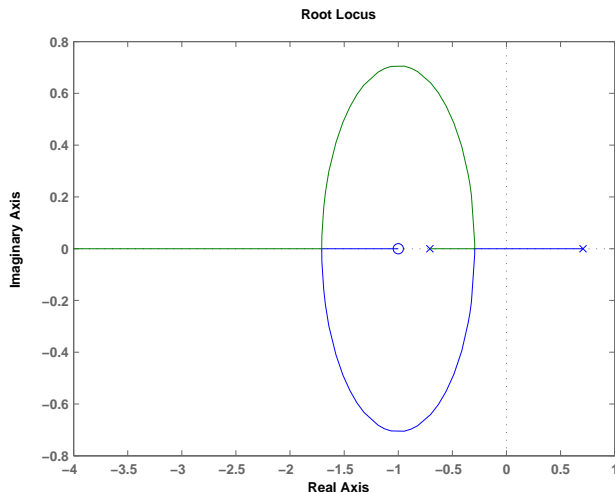
$$G(s) = \frac{1}{s^2 - \frac{1}{2}}$$



Root Locus on the Real Axis

Examples

The inverted pendulum with some derivative feedback: $\hat{K}(s) = k(1 + s)$



The Root Locus at High Gain

$k \rightarrow \infty$

Now let's look at what happens when gain increases.

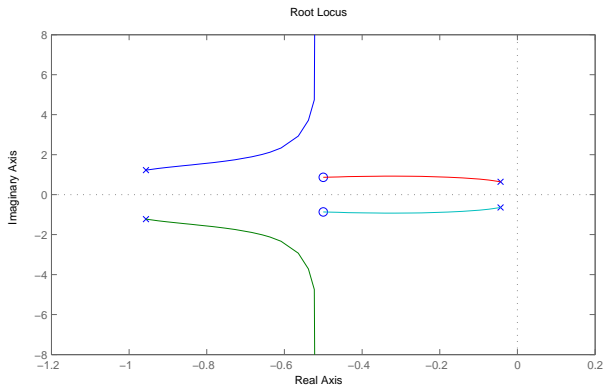
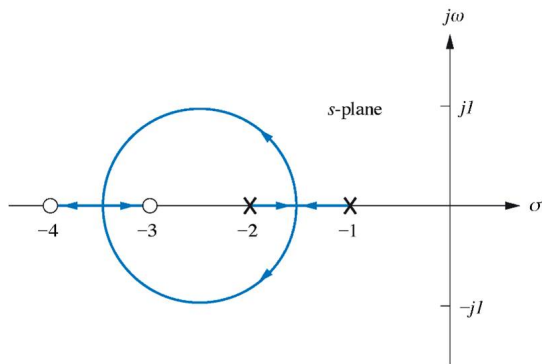


Figure: Suspension Problem

Conclusion: Some stable poles oscillate more.

The Root Locus at High Gain

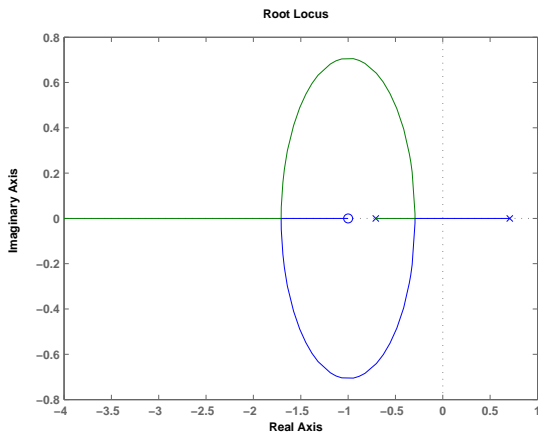
$$G(s) = \frac{(s + 3)(s + 4)}{(s + 1)(s + 2)}$$



Conclusion: Nothing Much Happens.

The Root Locus at High Gain

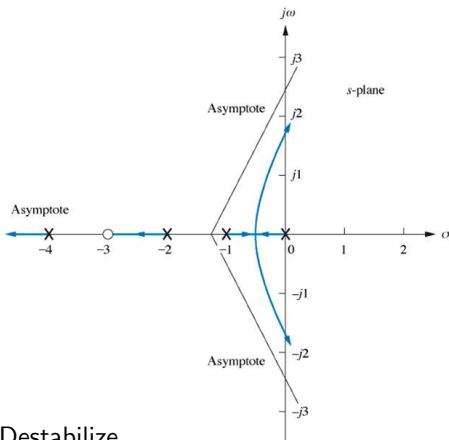
Again, Inverted Pendulum with derivative feedback.



Conclusion: Poles get **More Stable**.

The Root Locus at High Gain

$$G(s) = \frac{s + 3}{s(s + 1)(s + 2)(s + 4)}$$



Conclusion: Poles Destabilize.

Notice the **Asymptotes**.

The Root Locus at High Gain

So what happens when k is large?

Logically, there are **TWO CASES**:

- Poles can remain small.
- Poles can get big.

Consider the **Small Poles** ($\|s\| < \infty$).

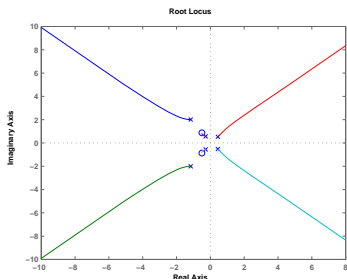
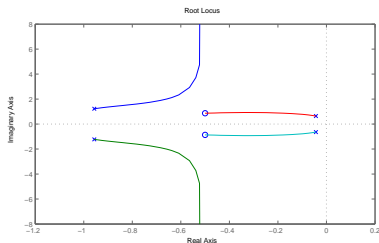
$$G(s) = \frac{n(s)}{d(s)}$$

- OL zeros are roots of $n(s) = 0$.
- OL poles are roots of $d(s) = 0$.

Now, closed loop:

$$\frac{kG(s)}{1 + kG(s)} = \frac{kn(s)}{d(s) + kn(s)}$$

CL poles are roots of
 $d(s) + kn(s) = 0$



The Root Locus at High Gain

At high gain, small CL poles are roots of

$$d(s) + kn(s) = 0$$

- If s is small, then $d(s)$ is small
- Hence as $k \rightarrow \infty$,

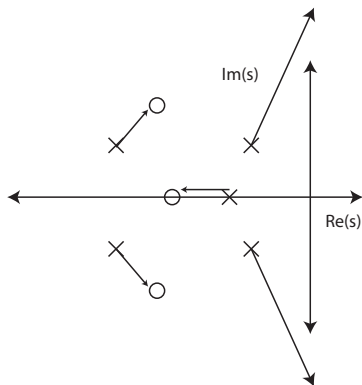
$$d(s) + kn(s) \cong kn(s)$$

As $k \rightarrow \infty$, small poles satisfy

$$d(s) + kn(s) \cong kn(s) = 0$$

which means $n(s) = 0!!!$

- $n(s) = 0$ means s is an OL zero!



At high gain, small CL poles are attracted by OL Zeros.

The Root Locus at High Gain

Asymptotics

Now consider the other possibility:

s also gets **Very Large**

$$d(s) + kn(s) = 0$$

In this case, $d(s)$ is not small.

Very Large solutions of

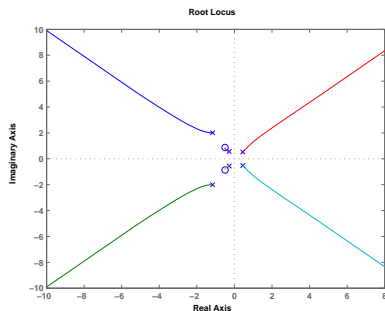
$$1 + kG(s)$$

are called asymptotics.

- asymptotics increase forever with k
 - ▶ $\lim_{k \rightarrow \infty} \|s\| = \infty$

Questions:

- Do asymptotes exist?
- Where do they go?



The Root Locus at High Gain

Asymptotics

Consider a point s which is **Very Large**

Recall that a point is on the root locus if

$$\angle G(s) = 180^\circ$$

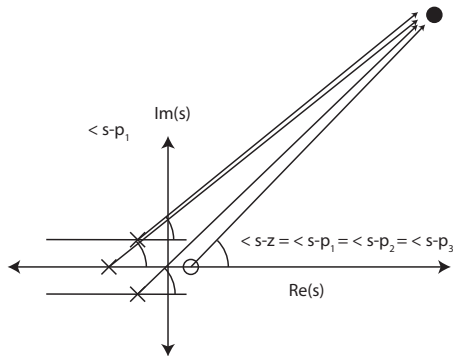
Which means:

$$\sum_{i=1}^m \angle(s - z_i) - \sum_{i=1}^n \angle(s - p_i) = -180^\circ$$

However, when $\|s\| \rightarrow \infty$,

All Angles are the Same!!!

$$\begin{aligned} \angle(s - z_1) &= \angle(s - z_2) = \dots = \angle(s - z_m) \\ &= \angle(s - p_1) = \angle(s - p_2) = \dots = \angle(s - p_n) = \angle_\infty \end{aligned}$$



Constructing the Root Locus

Asymptotics

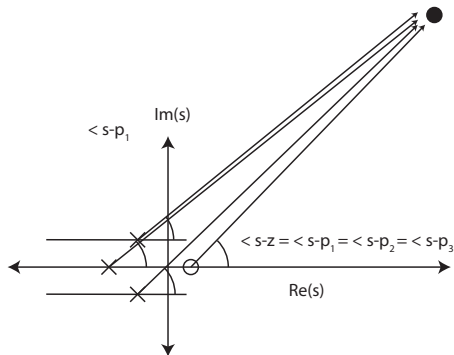
This makes life easier.

- just solve for one angle, \angle_{∞} .

$$\angle_{\infty} \cdot (m - n) = 180^{\circ}$$

where

- m is the number of OL zeros
- n is the number of OL poles



So asymptotics occur at

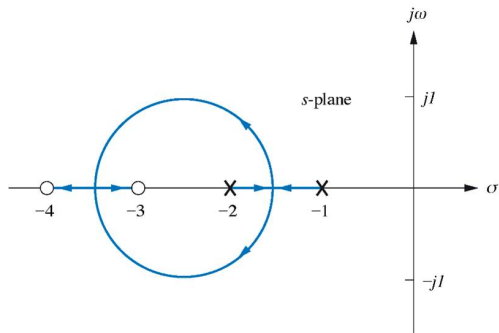
$$\angle_{\infty} = \frac{1}{m - n} (180^{\circ} + 360^{\circ}l)$$

for integers $l = 0, 1, 2, \dots$.

Asymptotes

Case 1: $n - m = 0$

$$G(s) = \frac{(s + 3)(s + 4)}{(s + 1)(s + 2)}$$



Count: 2 zeros, 2 poles.

$$m - n = 0 \quad \angle_{\infty} = \frac{1}{0} 180^{\circ} = \infty$$

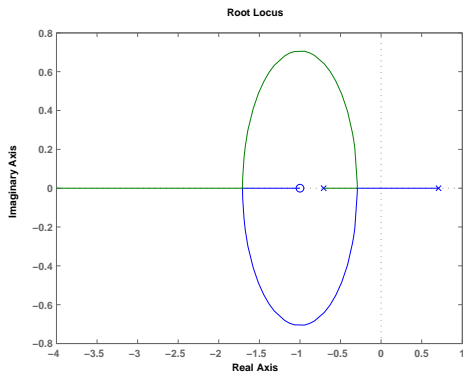
No Asymptotes.

Asymptotes

Case 2: $n - m = 1$

$$G(s) = \frac{1}{s^2 - \frac{1}{2}}$$

$$K(s) = k(1 + s)$$



Count: 1 zeros, 2 poles.

$$m - n = -1 \quad \angle_{\infty} = -180^{\circ}$$

1 asymptote at -180° .

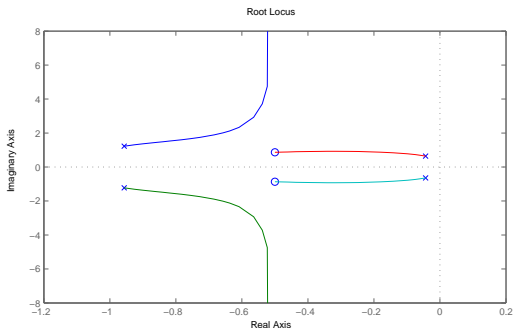
Asymptotes

Case 3: $n - m = 2$

The suspension system.

Count: 2 zeros, 4 poles.

$$m - n = -2$$



$$\angle_{\infty} = -\frac{1}{2} (180^{\circ} + 360^{\circ}l) = -90^{\circ} - 180^{\circ}l = -90^{\circ}, -270^{\circ}$$

2 vertical asymptotes at 90° and 270° .

Poles **MAY** destabilize at large gain.

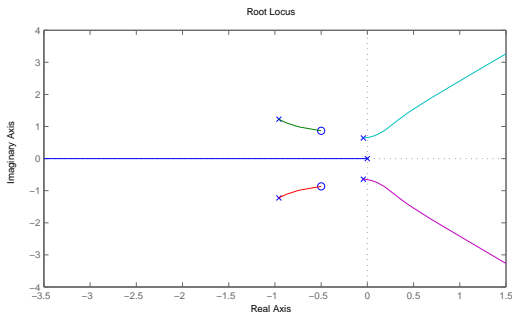
Asymptotes

Case 4: $n - m = 3$

The suspension system with integral feedback.

Count: 2 zeros, 5 poles.

$$m - n = -3$$



$$\angle_{\infty} = -\frac{1}{3} (180^{\circ} + 360^{\circ}l) = -60^{\circ} - 120^{\circ}l = -60^{\circ}, -180^{\circ}, -300^{\circ}$$

3 asymptotes at 60° , 180° and 300° .

Poles **WILL** destabilize at large gain.

Asymptotes

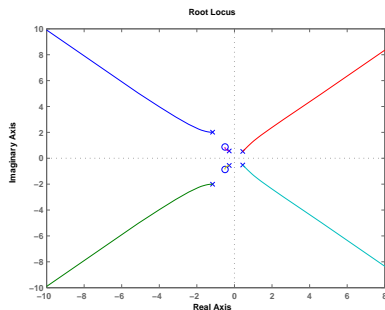
Case 5: $n - m = 4$

$n(s)$ is degree 2, $d(s)$ is degree 6.

$$G(s) = \frac{s^2 + s + 1}{s^6 + 2s^5 + 5s^4 - s^3 + 2s^2 + 1}$$

Count: 2 zeros, 6 poles.

$$m - n = -4$$



$$\angle_{\infty} = -\frac{1}{4} (180^{\circ} + 360^{\circ}l) = -45^{\circ} - 90^{\circ}l = -45^{\circ}, -135^{\circ}, -225^{\circ}, -315^{\circ}$$

4 asymptotes at 45° , 135° , 225° and 315° .

Poles **WILL** destabilize at large gain.

Asymptotics

Summary

Asymptotes depend only on relative number of poles and zeros.

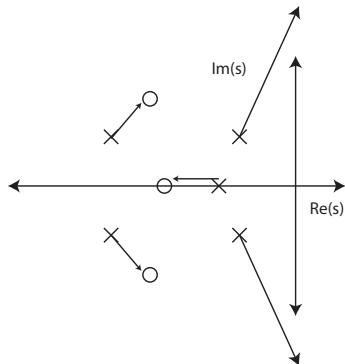
- Location of poles/zeros doesn't matter
 - ▶ At least not for the angle

One pole goes to each zero.

When there are more poles than zeros:

Cases:

- $n - m = 0$ - No Asymptotes
- $n - m = 1$ - Asymptote at 180°
- $n - m = 2$ - Asymptotes at $\pm 90^\circ$
- $n - m = 3$ - Asymptotes at $180^\circ, \pm 60^\circ$
- $n - m = 4$ - Asymptotes at $\pm 45^\circ$ and $\pm 135^\circ$



Summary

What have we learned today?

Review

- Definition of Root Locus

Points on the Real Axis

- Symmetry
- Drawing the Real Axis

What Happens at High Gain?

- The effect of Zeros
- Asymptotes

Next Lecture: Centers of Asymptotes, Break Points and Departure Angles