

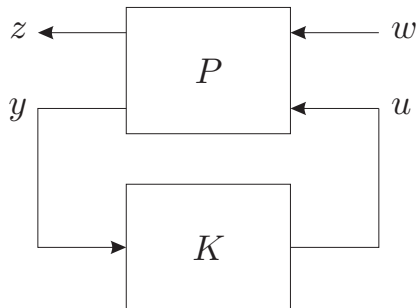
# LMI Methods in Optimal and Robust Control

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Lecture 10: An LMI for  $H_\infty$ -Optimal Output Feedback Control

# Optimal Output Feedback

Recall: Linear Fractional Transformation



**Plant:**

$$\begin{bmatrix} z \\ y \end{bmatrix} = \begin{bmatrix} P_{11} & P_{12} \\ P_{21} & P_{22} \end{bmatrix} \begin{bmatrix} w \\ u \end{bmatrix} \quad \text{where} \quad P = \left[ \begin{array}{c|cc} A & B_1 & B_2 \\ \hline C_1 & D_{11} & D_{12} \\ C_2 & D_{21} & D_{22} \end{array} \right]$$

**Controller:**

$$u = Ky \quad \text{where} \quad K = \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right]$$

# Defining the System Variables

Choose  $K$  to minimize

$$\|P_{11} + P_{12}(I - KP_{22})^{-1}KP_{21}\|$$

Equivalently choose  $\left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right]$  to minimize

$$\left\| \left[ \begin{array}{c|c} \left[ \begin{array}{cc} A & 0 \\ 0 & A_K \end{array} \right] + \left[ \begin{array}{cc} B_2 & 0 \\ 0 & B_K \end{array} \right] \left[ \begin{array}{cc} I & -D_K \\ -D_{22} & I \end{array} \right]^{-1} \left[ \begin{array}{cc} 0 & C_K \\ C_2 & 0 \end{array} \right] & \begin{array}{c} B_1 + B_2 D_K Q D_{21} \\ B_K Q D_{21} \end{array} \\ \hline \left[ \begin{array}{cc} C_1 & 0 \end{array} \right] + \left[ \begin{array}{cc} D_{12} & 0 \end{array} \right] \left[ \begin{array}{cc} I & -D_K \\ -D_{22} & I \end{array} \right]^{-1} \left[ \begin{array}{cc} 0 & C_K \\ C_2 & 0 \end{array} \right] & D_{11} + D_{12} D_K Q D_{21} \end{array} \right\|_{H_\infty}$$

where  $Q = (I - D_{22}D_K)^{-1}$ .

# Representing the Closed-Loop System

Recall the Matrix Inversion Lemma:

## Lemma 1.

$$\begin{aligned} & \begin{bmatrix} A & B \\ C & D \end{bmatrix}^{-1} \\ &= \begin{bmatrix} A^{-1} + A^{-1}B(D - CA^{-1}B)^{-1}CA^{-1} & -A^{-1}B(D - CA^{-1}B)^{-1} \\ -(D - CA^{-1}B)^{-1}CA^{-1} & (D - CA^{-1}B)^{-1} \end{bmatrix} \\ &= \begin{bmatrix} (A - BD^{-1}C)^{-1} & -(A - BD^{-1}C)^{-1}BD^{-1} \\ -D^{-1}C(A - BD^{-1}C)^{-1} & D^{-1} + D^{-1}C(A - BD^{-1}C)^{-1}BD^{-1} \end{bmatrix} \end{aligned}$$

# Closed Loop System is Nonlinear Function of $A_K, B_K, C_K, D_K$ .

Recall that

$$\begin{bmatrix} I & -D_K \\ -D_{22} & I \end{bmatrix}^{-1} = \begin{bmatrix} I + D_K Q D_{22} & D_K Q \\ Q D_{22} & Q \end{bmatrix}$$

where  $Q = (I - D_{22} D_K)^{-1}$ . Then

$$\begin{aligned} A_{cl} &:= \begin{bmatrix} A & 0 \\ 0 & A_K \end{bmatrix} + \begin{bmatrix} B_2 & 0 \\ 0 & B_K \end{bmatrix} \begin{bmatrix} I & -D_K \\ -D_{22} & I \end{bmatrix}^{-1} \begin{bmatrix} 0 & C_K \\ C_2 & 0 \end{bmatrix} \\ &= \begin{bmatrix} A & 0 \\ 0 & A_K \end{bmatrix} + \begin{bmatrix} B_2 & 0 \\ 0 & B_K \end{bmatrix} \begin{bmatrix} I + D_K Q D_{22} & D_K Q \\ Q D_{22} & Q \end{bmatrix} \begin{bmatrix} 0 & C_K \\ C_2 & 0 \end{bmatrix} \\ &= \begin{bmatrix} A + B_2 D_K Q C_2 & B_2 (I + D_K Q D_{22}) C_K \\ B_K Q C_2 & A_K + B_K Q D_{22} C_K \end{bmatrix} \end{aligned}$$

Likewise

$$\begin{aligned} C_{cl} &:= [C_1 \quad 0] + [D_{12} \quad 0] \begin{bmatrix} I + D_K Q D_{22} & D_K Q \\ Q D_{22} & Q \end{bmatrix} \begin{bmatrix} 0 & C_K \\ C_2 & 0 \end{bmatrix} \\ &= [C_1 + D_{12} D_K Q C_2 \quad D_{12} (I + D_K Q D_{22}) C_K] \end{aligned}$$

# A New Set of Decision Variables

Thus we have

$$\left[ \begin{array}{cc|c} A + B_2 D_K Q C_2 & B_2 (I + D_K Q D_{22}) C_K & B_1 + B_2 D_K Q D_{21} \\ B_K Q C_2 & A_K + B_K Q D_{22} C_K & B_K Q D_{21} \\ \hline C_1 + D_{12} D_K Q C_2 & D_{12} (I + D_K Q D_{22}) C_K & D_{11} + D_{12} D_K Q D_{21} \end{array} \right]$$

where  $Q = (I - D_{22} D_K)^{-1}$ .

- This is nonlinear in  $(A_K, B_K, C_K, D_K)$ .
- Hence we make a change of variables (First of several).

$$A_{K2} = A_K + B_K Q D_{22} C_K$$

$$B_{K2} = B_K Q$$

$$C_{K2} = (I + D_K Q D_{22}) C_K$$

$$D_{K2} = D_K Q$$

## A New parametrization of the closed-loop system.

This yields the system

$$\left[ \begin{array}{cc|c} \left[ \begin{array}{cc} A + B_2 D_{K2} C_2 & B_2 C_{K2} \\ B_{K2} C_2 & A_{K2} \end{array} \right] & & \begin{array}{c} B_1 + B_2 D_{K2} D_{21} \\ B_{K2} D_{21} \end{array} \\ \hline \left[ \begin{array}{cc} C_1 + D_{12} D_{K2} C_2 & D_{12} C_{K2} \end{array} \right] & & D_{11} + D_{12} D_{K2} D_{21} \end{array} \right]$$

Which is affine in  $\left[ \begin{array}{c|c} A_{K2} & B_{K2} \\ \hline C_{K2} & D_{K2} \end{array} \right]$ .

# Inverting the Variable Substitution

Recovering  $D_K$

Hence we can optimize over our new variables.

- System is linear in new variables and eliminates original variables.
- However, the change of variables must be **invertible**.

Now suppose we have  $D_{K2}$ . Then

$$D_{K2} = D_K Q = D_K (I - D_{22} D_K)^{-1}$$

implies that

$$D_K = D_{K2} (I - D_{22} D_K) = D_{K2} - D_{K2} D_{22} D_K$$

or

$$(I + D_{K2} D_{22}) D_K = D_{K2}$$

which can be inverted to get

$$D_K = (I + D_{K2} D_{22})^{-1} D_{K2}$$



# Inverting the Variable Substitution

Recovering  $C_K$

If we recall that

$$(I - QM)^{-1} = I + Q(I - MQ)^{-1}M$$

then we get

$$I + D_K Q D_{22} = I + D_K (I - D_{22} D_K)^{-1} D_{22} = (I - D_K D_{22})^{-1}$$

Examine the variable  $C_{K2}$

$$\begin{aligned} C_{K2} &= (I + D_K (I - D_{22} D_K)^{-1} D_{22}) C_K \\ &= (I - D_K D_{22})^{-1} C_K \end{aligned}$$

Hence, given  $D_K$  and  $C_{K2}$ , we can recover  $C_K$  as

$$C_K = (I - D_K D_{22}) C_{K2}$$

# Inverting the Variable Substitution

Once we have  $C_K$  and  $D_K$ , the other variables are easily recovered as

$$B_K = B_{K2}Q^{-1} = B_{K2}(I - D_{22}D_K)$$
$$A_K = A_{K2} - B_K(I - D_{22}D_K)^{-1}D_{22}C_K$$

To summarize, the original variables can be recovered as

$$D_K = (I + D_{K2}D_{22})^{-1}D_{K2}$$
$$B_K = B_{K2}(I - D_{22}D_K)$$
$$C_K = (I - D_KD_{22})C_{K2}$$
$$A_K = A_{K2} - B_K(I - D_{22}D_K)^{-1}D_{22}C_K$$

## Closed Loop System Parameters

$$\left[ \begin{array}{c|c} A_{cl} & B_{cl} \\ \hline C_{cl} & D_{cl} \end{array} \right] := \left[ \begin{array}{c|c} \left[ \begin{array}{cc} A + B_2 D_{K2} C_2 & B_2 C_{K2} \\ B_{K2} C_2 & A_{K2} \end{array} \right] & \begin{array}{c} B_1 + B_2 D_{K2} D_{21} \\ B_{K2} D_{21} \end{array} \\ \hline \left[ \begin{array}{cc} C_1 + D_{12} D_{K2} C_2 & D_{12} C_{K2} \end{array} \right] & D_{11} + D_{12} D_{K2} D_{21} \end{array} \right]$$

$$\left[ \begin{array}{cc} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{array} \right] = \left[ \begin{array}{ccc} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{array} \right] + \left[ \begin{array}{cc} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{array} \right] \left[ \begin{array}{cc} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{array} \right] \left[ \begin{array}{ccc} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{array} \right]$$

Or

$$A_{cl} = \left[ \begin{array}{cc} A & 0 \\ 0 & 0 \end{array} \right] + \left[ \begin{array}{cc} 0 & B_2 \\ I & 0 \end{array} \right] \left[ \begin{array}{cc} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{array} \right] \left[ \begin{array}{cc} 0 & I \\ C_2 & 0 \end{array} \right]$$

$$B_{cl} = \left[ \begin{array}{c} B_1 \\ 0 \end{array} \right] + \left[ \begin{array}{cc} 0 & B_2 \\ I & 0 \end{array} \right] \left[ \begin{array}{cc} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{array} \right] \left[ \begin{array}{c} 0 \\ D_{21} \end{array} \right]$$

$$C_{cl} = \left[ \begin{array}{cc} C_1 & 0 \end{array} \right] + \left[ \begin{array}{cc} 0 & D_{12} \end{array} \right] \left[ \begin{array}{cc} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{array} \right] \left[ \begin{array}{cc} 0 & I \\ C_2 & 0 \end{array} \right]$$

$$D_{cl} = \left[ \begin{array}{cc} D_{11} & 0 \end{array} \right] + \left[ \begin{array}{cc} 0 & D_{12} \end{array} \right] \left[ \begin{array}{cc} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{array} \right] \left[ \begin{array}{c} 0 \\ D_{21} \end{array} \right]$$

# Optimal Output Feedback Control

However, if we apply the KYP Lemma, the result is bilinear in  $X$  and  $A_K, B_K, C_K, D_K$

- Dual KYP Lemma is used for Controller Synthesis
- Primal KYP Lemma is used for observer Synthesis
- For Observer-Based Controller Synthesis, we need both Primal AND Dual forms....

## Lemma 2 (Transformation Lemma).

*Suppose that*

$$\begin{bmatrix} Y_1 & I \\ I & X_1 \end{bmatrix} > 0$$

*Then there exist  $X_2, X_3, Y_2, Y_3$  such that*

$$X = \begin{bmatrix} X_1 & X_2 \\ X_2^T & X_3 \end{bmatrix} = \begin{bmatrix} Y_1 & Y_2 \\ Y_2^T & Y_3 \end{bmatrix}^{-1} = Y^{-1} > 0$$

*where  $Y_{cl} = \begin{bmatrix} Y_1 & I \\ Y_2^T & 0 \end{bmatrix}$  has full rank.*

# Optimal Output Feedback Control

However, if we apply the KYP Lemma, the result is bilinear in  $X$  and

$A_d, B_d, C_d, D_d$ :

- Dual KYP Lemma is used for Controller Synthesis
- Primal KYP Lemma is used for observer Synthesis
- For Observer-Based Controller Synthesis, we need both Primal AND Dual forms...

### Lemma 2 (Transformation Lemma).

Suppose that

$$\begin{bmatrix} Y_1 & I \\ I & X_3 \end{bmatrix} > 0$$

Then there exist  $X_2, X_1, Y_2, Y_3$  such that

$$X = \begin{bmatrix} X_1 & X_2 \\ X_2^T & X_3 \end{bmatrix} = \begin{bmatrix} Y_1 & Y_2 \\ Y_2^T & Y_3 \end{bmatrix}^{-1} = Y^{-1} > 0$$

where  $Y_d = \begin{bmatrix} Y_1 & I \\ Y_2^T & 0 \end{bmatrix}$  has full rank.

The primal variable is

$$\begin{bmatrix} X_1 & X_2 \\ X_2^T & X_3 \end{bmatrix}$$

The dual variable is

$$\begin{bmatrix} Y_1 & Y_2 \\ Y_2^T & Y_3 \end{bmatrix}$$

## Proof.

- Since

$$\begin{bmatrix} Y_1 & I \\ I & X_1 \end{bmatrix} > 0,$$

by the Schur complement  $X_1 > 0$  and  $X_1^{-1} - Y_1 < 0$ . Since  $I - X_1 Y_1 = X_1(X_1^{-1} - Y_1)$ , we conclude that  $I - X_1 Y_1$  is invertible.

- Choose any two square invertible matrices  $X_2$  and  $Y_2$  such that

$$X_2 Y_2^T = I - X_1 Y_1$$

- Because  $X_2$  and  $Y_2$  are invertible,

$$Y_{cl}^T = \begin{bmatrix} Y_1 & Y_2 \\ I & 0 \end{bmatrix} \text{ and } X_{cl} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix}$$

are also non-singular.



# Optimal Output Feedback Control

## Proof.

• Since

$$\begin{bmatrix} Y_1 & I \\ I & X_1 \end{bmatrix} > 0,$$

by the Schur complement  $X_1 > 0$  and  $X_1^{-1} - Y_1 < 0$ . Since  $I - X_1 Y_1 = X_1 (X_1^{-1} - Y_1)$ , we conclude that  $I - X_1 Y_1$  is invertible.

• Choose any two square invertible matrices  $X_2$  and  $Y_2$  such that

$$X_2 Y_2^T = I - X_1 Y_1$$

• Because  $X_1$  and  $Y_1$  are invertible,

$$Y_2^T = \begin{bmatrix} Y_1 & Y_2 \\ I & 0 \end{bmatrix} \text{ and } X_2 = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix}$$

are also non-singular.

$$\begin{bmatrix} Y_{cl}^T & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix}$$

will be the half-dual transformation.

- $Y_{cl}^T$  contains the top half of the dual Lyapunov variable.
- $X_{cl}$  contains the bottom half of the primal Lyapunov variable.

# Optimal Output Feedback Control

Proof.

- Now define  $X$  and  $Y$  as

$$X = Y_{cl}^{-T} X_{cl} \quad \text{and} \quad Y = X_{cl}^{-1} Y_{cl}^T.$$

Then

$$XY = Y_{cl}^{-1} X_{cl} X_{cl}^{-1} Y_{cl} = I$$

□

Note that

$$Y_{cl}^T X = X_{cl} \quad \text{and} \quad X_{cl} Y = Y_{cl}^T$$



# Optimal Output Feedback Control

Proof.

\* Now define  $X$  and  $Y$  as

$$X = Y_2^{-T} X_{cl} \quad \text{and} \quad Y = X_2^{-1} Y_2^T.$$

Then

$$XY = Y_2^{-1} X_{cl} X_2^{-1} Y_2^T = I$$

Note that

$$Y_2^T X = X_{cl} \quad \text{and} \quad X_{cl} Y = Y_2^T$$

We will be applying the half-dual transformation

$$\begin{aligned}
 & \begin{bmatrix} Y_{cl}^T & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \overbrace{\begin{bmatrix} A^T X + X A & X B & C^T \\ B^T X & -\gamma I & D^T \\ C & D & -\gamma I \end{bmatrix}}^{\text{primal KYP lemma}} \begin{bmatrix} Y_{cl} & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \\
 &= \begin{bmatrix} Y_{cl}^T A^T X Y_{cl} + Y_{cl}^T X A Y_{cl} & Y_{cl}^T X B & Y_{cl}^T C^T \\ B^T X Y_{cl} & -\gamma I & D^T \\ C Y_{cl} & D & -\gamma I \end{bmatrix} \\
 &= \begin{bmatrix} Y_{cl}^T A^T X_{cl}^T + X_{cl} A Y_{cl} & X_{cl} B & Y_{cl}^T C^T \\ B^T X_{cl}^T & -\gamma I & D^T \\ C Y_{cl} & D & -\gamma I \end{bmatrix}
 \end{aligned}$$

Since

$$Y_{cl}^T = \begin{bmatrix} Y_1 & Y_2 \\ I & 0 \end{bmatrix} \quad \text{and} \quad X_{cl} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix},$$

this will only work if  $Y_2$  and  $X_2$  are somehow eliminated from the expression.

## Lemma 3 (Converse Transformation Lemma).

Given  $X = \begin{bmatrix} X_1 & X_2 \\ X_2^T & X_3 \end{bmatrix} > 0$  where  $X_2$  has full column rank. Let

$$X^{-1} = Y = \begin{bmatrix} Y_1 & Y_2 \\ Y_2^T & Y_3 \end{bmatrix}$$

then

$$\begin{bmatrix} Y_1 & I \\ I & X_1 \end{bmatrix} > 0$$

and  $Y_{cl} = \begin{bmatrix} Y_1 & I \\ Y_2^T & 0 \end{bmatrix}$  has full column rank.

This result shows the Transformation Lemma is not conservative.

# Optimal Output Feedback Control

## Proof.

Since  $X_2$  is full rank,  $X_{cl} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix}$  also has full column rank. Note that  $YX = I$  implies

$$Y_{cl}^T X = \begin{bmatrix} Y_1 & Y_2 \\ I & 0 \end{bmatrix} \begin{bmatrix} X_1 & X_2 \\ X_2^T & X_3 \end{bmatrix} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix} = X_{cl}.$$

Hence

$$Y_{cl}^T = \begin{bmatrix} Y_1 & Y_2 \\ I & 0 \end{bmatrix} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix} Y = X_{cl} Y$$

has full column rank. Now, since  $XY = I$  implies  $X_1 Y_1 + X_2 Y_2^T = I$ , we have

$$X_{cl} Y_{cl} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix} \begin{bmatrix} Y_1 & I \\ Y_2^T & 0 \end{bmatrix} = \begin{bmatrix} Y_1 & I \\ X_1 Y_1 + X_2 Y_2^T & X_1 \end{bmatrix} = \begin{bmatrix} Y_1 & I \\ I & X_1 \end{bmatrix}$$

Furthermore, because  $Y_{cl}$  has full rank,

$$\begin{bmatrix} Y_1 & I \\ I & X_1 \end{bmatrix} = X_{cl} Y_{cl} = X_{cl} Y X_{cl}^T = Y_{cl}^T X Y_{cl} > 0$$

# Optimal Output Feedback Control

## Proof.

Since  $X_2$  is full rank,  $X_{cl} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix}$  also has full column rank. Note that  $YX = I$  implies

$$Y_2^T X = \begin{bmatrix} Y_1 & Y_2 \\ I & 0 \end{bmatrix} \begin{bmatrix} X_1 & X_2 \\ X_1^T & X_2^T \end{bmatrix} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix} = X_{cl}$$

Hence

$$Y_2^T = \begin{bmatrix} Y_1 & Y_2 \\ I & 0 \end{bmatrix} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix} Y = X_{cl} Y$$

has full column rank. Now, since  $XV = I$  implies  $X_1 Y_1 + X_2 Y_2^T = I$ , we have

$$X_{cl} Y_{cl} = \begin{bmatrix} I & 0 \\ X_1 & X_2 \end{bmatrix} \begin{bmatrix} Y_1 & I \\ Y_2^T & 0 \end{bmatrix} = \begin{bmatrix} X_1 Y_1 + X_2 Y_2^T & I \\ I & X_2 \end{bmatrix} = \begin{bmatrix} Y_1 & I \\ I & X_2 \end{bmatrix}$$

Furthermore, because  $Y_{cl}$  has full rank,

$$\begin{bmatrix} Y_1 & I \\ I & X_2 \end{bmatrix} = X_{cl} Y_{cl} = X_{cl} Y X_{cl}^T = Y_2^T X Y_{cl} > 0$$

Note the relationship:

$$\begin{bmatrix} Y_1 & I \\ I & X_2 \end{bmatrix} = X_{cl} Y_{cl}$$

Note how both  $X_2$  and  $Y_2$  vanish?

# Optimal Output Feedback Control

## Theorem 4.

The following are equivalent.

- There exists a  $K = \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right]$  such that  $\|\underline{S}(K, P)\|_{H_\infty} < \gamma$ .

- There exist  $X_1, Y_1, A_n, B_n, C_n, D_n$  such that  $\begin{bmatrix} X_1 & I \\ I & Y_1 \end{bmatrix} > 0$

$$\begin{bmatrix} AY_1 + Y_1A^T + B_2C_n + C_n^T B_2^T & *^T & *^T & *^T \\ A^T + A_n + [B_2D_nC_2]^T & X_1A + A^T X_1 + B_nC_2 + C_2^T B_n^T & *^T & *^T \\ [B_1 + B_2D_nD_{21}]^T & [X_1B_1 + B_nD_{21}]^T & -\gamma I & \\ C_1Y_1 + D_{12}C_n & C_1 + D_{12}D_nC_2 & D_{11} + D_{12}D_nD_{21} & -\gamma I \end{bmatrix} < 0$$

Moreover, such a controller is given by  $D_K = (I + D_{K2}D_{22})^{-1}D_{K2}$ ,

$B_K = B_{K2}(I - D_{22}D_K)$ ,  $C_K = (I - D_KD_{22})C_{K2}$ ,

$A_K = A_{K2} - B_K(I - D_{22}D_K)^{-1}D_{22}C_K$  where

$$\left[ \begin{array}{c|c} A_{K2} & B_{K2} \\ \hline C_{K2} & D_{K2} \end{array} \right] = \begin{bmatrix} X_2 & X_1B_2 \\ 0 & I \end{bmatrix}^{-1} \left[ \begin{array}{c|c} A_n & B_n \\ \hline C_n & D_n \end{array} \right] - \begin{bmatrix} X_1AY_1 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} Y_2^T & 0 \\ C_2Y_1 & I \end{bmatrix}^{-1}$$

for any full-rank  $X_2$  and  $Y_2$  such that

$$\begin{bmatrix} X_1 & X_2 \\ X_2^T & X_3 \end{bmatrix} = \begin{bmatrix} Y_1 & Y_2 \\ Y_2^T & Y_3 \end{bmatrix}^{-1}$$

# Optimal Output Feedback Control

## Proof: If.

Suppose there exist  $X_1, Y_1, A_n, B_n, C_n, D_n$  such that the LMI is feasible. Since

$$\begin{bmatrix} X_1 & I \\ I & Y_1 \end{bmatrix} > 0,$$

by the transformation lemma, there exist  $X_2, X_3, Y_2, Y_3$  such that

$$X := \begin{bmatrix} X & X_2 \\ X_2^T & X_3 \end{bmatrix} = \begin{bmatrix} Y & Y_2 \\ Y_2^T & Y_3 \end{bmatrix}^{-1} > 0$$

where  $Y_{cl} = \begin{bmatrix} Y & I \\ Y_2^T & 0 \end{bmatrix}$  has full row rank. Let  $K = \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right]$  where

$$D_K = (I + D_{K2}D_{22})^{-1}D_{K2}$$

$$B_K = B_{K2}(I - D_{22}D_K)$$

$$C_K = (I - D_KD_{22})C_{K2}$$

$$A_K = A_{K2} - B_K(I - D_{22}D_K)^{-1}D_{22}C_{K2}.$$

# Optimal Output Feedback Control

## Proof: If.

and where

$$\begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} = \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix}^{-1} \left[ \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} - \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix} \right] \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix}^{-1}.$$

Thus  $\underline{S} \left( \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right], P \right) = \left[ \begin{array}{c|c} A_{cl} & B_{cl} \\ \hline C_{cl} & D_{cl} \end{array} \right]$ , where

$$\begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} = \begin{bmatrix} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{bmatrix} + \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix}$$

$$= \begin{bmatrix} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{bmatrix} + \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix}$$

$$\begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix}^{-1} \left[ \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} - \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix} \right] \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix}^{-1} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix}$$

We now apply the KYP lemma to this system.

# Optimal Output Feedback Control

## Proof: If.

Since  $Y_{cl}$  is full rank,  $\|\underline{S}(K, P)\|_{H_\infty} < \gamma$  if and only if

$$\begin{bmatrix} Y_{cl}^T & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} A_{cl}^T X + X A_{cl} & X B_{cl} & C_{cl}^T \\ B_{cl}^T X & -\gamma I & D_{cl}^T \\ C_{cl} & D_{cl} & -\gamma I \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} < 0$$

In the next several slides, we show that

$$\begin{bmatrix} Y_{cl}^T & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} A_{cl}^T X + X A_{cl} & X B_{cl} & C_{cl}^T \\ B_{cl}^T X & -\gamma I & D_{cl}^T \\ C_{cl} & D_{cl} & -\gamma I \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} =$$

$$\begin{bmatrix} AY_1 + Y_1 A^T + B_2 C_n + C_n^T B_2^T & *^T & *^T & *^T \\ A^T + A_n + [B_2 D_n C_2]^T & X_1 A + A^T X_1 + B_n C_2 + C_2^T B_n^T & *^T & *^T \\ [B_1 + B_2 D_n D_{21}]^T & [X_1 B_1 + B_n D_{21}]^T & -\gamma I & *^T \\ C_1 Y_1 + D_{12} C_n & C_1 + D_{12} D_n C_2 & D_{11} + D_{12} D_n D_{21} & -\gamma I \end{bmatrix} < 0$$

□



# Optimal Output Feedback Control

## Proof: If.

First we note that

$$\begin{aligned} & \begin{bmatrix} Y_{cl}^T & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} A_{cl}^T X + X A_{cl} & X B_{cl} & C_{cl}^T \\ B_{cl}^T X_{cl} & -\gamma I & D_{cl}^T \\ C_{cl} & D_{cl} & -\gamma I \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \\ &= \begin{bmatrix} Y_{cl}^T A_{cl}^T X_{cl}^T + X_{cl} A_{cl} Y_{cl} & X_{cl} B_{cl} & Y_{cl}^T C_{cl}^T \\ B_{cl}^T X_{cl}^T & -\gamma I & D_{cl}^T \\ C_{cl} Y_{cl} & D_{cl} & -\gamma I \end{bmatrix} \end{aligned}$$

By comparing terms, our goal is then to show that

$$\begin{aligned} & \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} = \begin{bmatrix} X_{cl} A_{cl} Y_{cl} & X_{cl} B_{cl} \\ C_{cl} Y_{cl} & D_{cl} \end{bmatrix} \\ &= \begin{bmatrix} A Y_1 + B_2 C_n & A + B_2 D_n C_2 & B_1 + B_2 D_n D_{21} \\ A_n & X_1 A + B_n C_2 & X_1 + B_n D_{21} \\ C_1 Y_1 + D_{12} D_n & C_1 + D_{12} D_n C_2 & D_{11} + D_{12} D_n D_{21} \end{bmatrix} \end{aligned}$$



# Optimal Output Feedback Control

**Proof: If.**

$$\begin{aligned} & \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\ &= \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \left( \begin{bmatrix} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{bmatrix} + \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix} \right) \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\ &= \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\ & \quad + \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \end{aligned}$$

□

We now examine each of these terms separately.

## Proof: If.

First, we have

$$\begin{aligned} & \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\ &= \begin{bmatrix} I & 0 & 0 \\ X_1 & X_2 & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{bmatrix} \begin{bmatrix} Y_1 & I & 0 \\ Y_2^T & 0 & 0 \\ 0 & 0 & I \end{bmatrix} \\ &= \begin{bmatrix} I & 0 & 0 \\ X_1 & X_2 & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} AY_1 & A & B_1 \\ 0 & 0 & 0 \\ C_1Y_1 & C_1 & D_{11} \end{bmatrix} \\ &= \begin{bmatrix} AY_1 & A & B_1 \\ X_1AY_1 & X_1A & X_1B_1 \\ C_1Y_1 & C_1 & D_{11} \end{bmatrix} \end{aligned}$$



# Optimal Output Feedback Control

## Proof: If.

Next, we observe that

$$\begin{aligned} & \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\ &= \begin{bmatrix} I & 0 & 0 \\ X_1 & X_2 & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix} \begin{bmatrix} Y_1 & I & 0 \\ Y_2^T & 0 & 0 \\ 0 & 0 & I \end{bmatrix} \\ &= \begin{bmatrix} 0 & B_2 \\ X_2 & X_1 B_1 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} Y_2^T & 0 & 0 \\ C_2 Y_1 & C_2 & D_{21} \end{bmatrix} \\ &= \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix} \begin{bmatrix} I & 0 & 0 \\ 0 & C_2 & D_{21} \end{bmatrix} \end{aligned}$$

The obvious change of variables is now

$$\begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} = \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix} + \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix}.$$

## Proof: If.

Next, we note that by definition

$$\begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} = \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix}^{-1} \left[ \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} - \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix} \right] \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix}^{-1}.$$

and hence

$$\begin{aligned} & \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix} \\ &= \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix} \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix}^{-1} \left[ \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} - \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix} \right] \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix}^{-1} \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix} \\ &= \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} - \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix} \end{aligned}$$

□

## Proof: If.

We conclude that

$$\begin{aligned}
 & \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\
 &= \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix} \begin{bmatrix} I & 0 & 0 \\ 0 & C_2 & D_{21} \end{bmatrix} \\
 &= \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \left( \begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} - \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix} \right) \begin{bmatrix} I & 0 & 0 \\ 0 & C_2 & D_{21} \end{bmatrix} \\
 &= \begin{bmatrix} B_2 C_n & B_2 D_n C_2 & B_2 D_n D_{21} \\ A_n - X_1 A Y_1 & B_n C_2 & B_n D_{21} \\ D_{12} C_n & D_{12} D_n C_2 & D_{12} D_n D_{21} \end{bmatrix}
 \end{aligned}$$



# Optimal Output Feedback Control

## Proof: If.

In summary, we have

$$\begin{aligned} & \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{cl} & B_{cl} \\ C_{cl} & D_{cl} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} = \begin{bmatrix} X_{cl}A_{cl}Y_{cl} & X_{cl}B_{cl} \\ C_{cl}Y_{cl} & D_{cl} \end{bmatrix} \\ & = \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} A & 0 & B_1 \\ 0 & 0 & 0 \\ C_1 & 0 & D_{11} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\ & \quad + \begin{bmatrix} X_{cl} & 0 \\ 0 & I \end{bmatrix} \begin{bmatrix} 0 & B_2 \\ I & 0 \\ 0 & D_{12} \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} 0 & I & 0 \\ C_2 & 0 & D_{21} \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 \\ 0 & I \end{bmatrix} \\ & = \begin{bmatrix} AY_1 & A & B_1 \\ X_1AY_1 & X_1A & X_1B_1 \\ C_1Y_1 & C_1 & D_{11} \end{bmatrix} + \begin{bmatrix} B_2C_n & B_2D_nC_2 & B_2D_nD_{21} \\ A_n - X_1AY_1 & B_nC_2 & B_nD_{21} \\ D_{12}C_n & D_{12}D_nC_2 & D_{12}D_nD_{21} \end{bmatrix} \\ & = \begin{bmatrix} AY_1 + B_2C_n & A + B_2D_nC_2 & B_1 + B_2D_nD_{21} \\ A_n & X_1A + B_nC_2 & X_1 + B_nD_{21} \\ C_1Y_1 + D_{12}D_n & C_1 + D_{12}D_nC_2 & D_{11} + D_{12}D_nD_{21} \end{bmatrix} \end{aligned}$$

as desired. □

# Optimal Output Feedback Control

## Proof: If.

We therefore conclude that

$$\left\| \underline{S} \left( \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right], P \right) \right\|_{H_\infty} = \left\| \left[ \begin{array}{c|c} A_{cl} & B_{cl} \\ \hline C_{cl} & D_{cl} \end{array} \right] \right\|_{H_\infty} < \gamma$$

since

$$\begin{aligned} & \begin{bmatrix} Y_{cl}^T & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} A_{cl}^T X + X A_{cl} & X B_{cl} & C_{cl}^T \\ B_{cl}^T X & -\gamma I & D_{cl}^T \\ C_{cl} & D_{cl} & -\gamma I \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \\ &= \begin{bmatrix} Y_{cl}^T A_{cl}^T X_{cl}^T + X_{cl} A_{cl} Y_{cl} & X_{cl} B_{cl} & Y_{cl}^T C_{cl}^T \\ B_{cl}^T X_{cl}^T & -\gamma I & D_{cl}^T \\ C_{cl} Y_{cl} & D_{cl} & -\gamma I \end{bmatrix} = \\ & \begin{bmatrix} AY_1 + Y_1 A^T + B_2 C_n + C_n^T B_2^T & *^T & *^T & *^T \\ A^T + A_n + [B_2 D_n C_2]^T & X_1 A + A^T X_1 + B_n C_2 + C_2^T B_n^T & *^T & *^T \\ [B_1 + B_2 D_n D_{21}]^T & [X_1 B_1 + B_n D_{21}]^T & -\gamma I & *^T \\ C_1 Y_1 + D_{12} C_n & C_1 + D_{12} D_n C_2 & D_{11} + D_{12} D_n D_{21} & -\gamma I \end{bmatrix} < 0 \end{aligned}$$



# Optimal Output Feedback Control

## Proof: Only If.

Likewise, if

$$\left\| \underline{S} \left( \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right], P \right) \right\|_{H_\infty} = \left\| \left[ \begin{array}{c|c} A_{cl} & B_{cl} \\ \hline C_{cl} & D_{cl} \end{array} \right] \right\|_{H_\infty} < \gamma$$

then there exists  $X > 0$  such that

$$\begin{bmatrix} A_{cl}^T X + X A_{cl} & X B_{cl} & C_{cl}^T \\ B_{cl}^T X & -\gamma I & D_{cl}^T \\ C_{cl} & D_{cl} & -\gamma I \end{bmatrix} < 0.$$

Define

$$Y = \begin{bmatrix} Y_1 & Y_2 \\ Y_2^T & Y_3 \end{bmatrix} = \begin{bmatrix} X_1 & X_2 \\ X_2^T & X_3 \end{bmatrix}^{-1} = X^{-1} \quad \text{and} \quad Y_{cl} = \begin{bmatrix} Y_1 & I \\ Y_2^T & 0 \end{bmatrix}$$

where because the inequalities are strict, we can assume that  $X_2$  has full row rank. Then, according to the converse transformation lemma,  $Y_{cl}$  has full row rank and

$$\begin{bmatrix} X_1 & I \\ I & Y_1 \end{bmatrix} > 0.$$

# Optimal Output Feedback Control

## Proof: Only If.

Now define

$$\begin{bmatrix} A_n & B_n \\ C_n & D_n \end{bmatrix} = \begin{bmatrix} X_2 & X_1 B_2 \\ 0 & I \end{bmatrix} \begin{bmatrix} A_{K2} & B_{K2} \\ C_{K2} & D_{K2} \end{bmatrix} \begin{bmatrix} Y_2^T & 0 \\ C_2 Y_1 & I \end{bmatrix} + \begin{bmatrix} X_1 A Y_1 & 0 \\ 0 & 0 \end{bmatrix}$$

where

$$A_{K2} = A_K + B_K(I - D_{22}D_K)^{-1}D_{22}C_K \quad B_{K2} = B_K(I - D_{22}D_K)^{-1}$$

$$C_{K2} = (I + D_K(I - D_{22}D_K)^{-1}D_{22})C_K \quad D_{K2} = D_K(I - D_{22}D_K)^{-1}$$

and we have

$$\begin{bmatrix} AY_1 + Y_1 A^T + B_2 C_n + C_n^T B_2^T & *^T & *^T & *^T \\ A^T + A_n + [B_2 D_n C_2]^T & X_1 A + A^T X_1 + B_n C_2 + C_2^T B_n^T & *^T & *^T \\ [B_1 + B_2 D_n D_{21}]^T & [X_1 B_1 + B_n D_{21}]^T & -\gamma I & *^T \\ C_1 Y_1 + D_{12} C_n & C_1 + D_{12} D_n C_2 & D_{11} + D_{12} D_n D_{21} & -\gamma I \end{bmatrix}$$

$$= \begin{bmatrix} Y_{cl}^T A_{cl}^T X_{cl}^T + X_{cl} A_{cl} Y_{cl} & X_{cl} B_{cl} & Y_{cl}^T C_{cl}^T \\ B_{cl}^T X_{cl}^T & -\gamma I & D_{cl}^T \\ C_{cl} Y_{cl} & D_{cl} & -\gamma I \end{bmatrix}$$

$$= \begin{bmatrix} Y_{cl}^T & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} A_{cl}^T X + X A_{cl} & X B_{cl} & C_{cl}^T \\ B_{cl}^T X & -\gamma I & D_{cl}^T \\ C_{cl} & D_{cl} & -\gamma I \end{bmatrix} \begin{bmatrix} Y_{cl} & 0 & 0 \\ 0 & I & 0 \\ 0 & 0 & I \end{bmatrix} < 0. \square$$

# Conclusion

To solve the  $H_\infty$ -optimal state-feedback problem, we solve

$\min_{\gamma, X_1, Y_1, A_n, B_n, C_n, D_n} \gamma$  such that

$$\begin{bmatrix} X_1 & I \\ I & Y_1 \end{bmatrix} > 0$$

$$\begin{bmatrix} AY_1 + Y_1 A^T + B_2 C_n + C_n^T B_2^T & *^T & *^T & *^T \\ A^T + A_n + [B_2 D_n C_2]^T & X_1 A + A^T X_1 + B_n C_2 + C_2^T B_n^T & *^T & *^T \\ [B_1 + B_2 D_n D_{21}]^T & [X_1 B_1 + B_n D_{21}]^T & -\gamma I & *^T \\ C_1 Y_1 + D_{12} C_n & C_1 + D_{12} D_n C_2 & D_{11} + D_{12} D_n D_{21} & -\gamma I \end{bmatrix} < 0$$

# Conclusion

Then, we construct our controller using

$$D_K = (I + D_{K2}D_{22})^{-1}D_{K2}$$

$$B_K = B_{K2}(I - D_{22}D_K)$$

$$C_K = (I - D_KD_{22})C_{K2}$$

$$A_K = A_{K2} - B_K(I - D_{22}D_K)^{-1}D_{22}C_K.$$

where

$$\left[ \begin{array}{c|c} A_{K2} & B_{K2} \\ \hline C_{K2} & D_{K2} \end{array} \right] = \left[ \begin{array}{cc} X_2 & X_1B_2 \\ 0 & I \end{array} \right]^{-1} \left[ \left[ \begin{array}{cc} A_n & B_n \\ \hline C_n & D_n \end{array} \right] - \left[ \begin{array}{cc} X_1AY_1 & 0 \\ 0 & 0 \end{array} \right] \right] \left[ \begin{array}{cc} Y_2^T & 0 \\ \hline C_2Y_1 & I \end{array} \right]^{-1}.$$

and where  $X_2$  and  $Y_2$  are any matrices which satisfy  $X_2Y_2^T = I - X_1Y_1$ .

- e.g. Let  $Y_2 = I$  and  $X_2 = I - X_1Y_1$ .
- The optimal controller is NOT uniquely defined.
- Don't forget to check invertibility of  $I - D_{22}D_K$

# Conclusion

The  $H_\infty$ -optimal controller is a dynamic system.

- Transfer Function  $\hat{K}(s) = \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right]$

Minimizes the effect of external input ( $w$ ) on external output ( $z$ ).

$$\|z\|_{L_2} \leq \|\underline{S}(P, K)\|_{H_\infty} \|w\|_{L_2}$$

- Minimum Energy Gain

# Alternative Formulation

## Theorem 5.

Let  $N_o$  and  $N_c$  be full-rank matrices whose images satisfy

$$\text{Im } N_o = \text{Ker } \begin{bmatrix} C_2 & D_{21} \end{bmatrix}$$

$$\text{Im } N_c = \text{Ker } \begin{bmatrix} B_2^T & D_{12}^T \end{bmatrix}$$

Then the following are equivalent.

- There exists a  $\hat{K} = \left[ \begin{array}{c|c} A_K & B_K \\ \hline C_K & D_K \end{array} \right]$  such that  $\|S(K, P)\|_{H_\infty} < 1$ .
- There exist  $X_1, Y_1$  such that  $\begin{bmatrix} X_1 & I \\ I & Y_1 \end{bmatrix} > 0$  and

$$\begin{bmatrix} N_o & 0 \\ 0 & I \end{bmatrix}^T \begin{bmatrix} X_1 A + A^T X_1 & *^T & *^T \\ B_1^T X_1 & -I & *^T \\ C_1 & D_{11} & -I \end{bmatrix} \begin{bmatrix} N_o & 0 \\ 0 & I \end{bmatrix} < 0$$

$$\begin{bmatrix} N_c & 0 \\ 0 & I \end{bmatrix}^T \begin{bmatrix} A Y_1 + Y_1 A & *^T & *^T \\ C_1 Y_1 & -I & *^T \\ B_1^T & D_{11}^T & -I \end{bmatrix} \begin{bmatrix} N_c & 0 \\ 0 & I \end{bmatrix} < 0$$