

Systems Analysis and Control

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Lecture 16: Generalized Root Locus and Controller Design

In this Lecture, you will learn:

Generalized Root Locus?

- What about changing *OTHER* parameters
- T_D , T_I , et c.
- mass, damping, et c.

Compensation via Root-Locus

- Introduction
- Pole-Zero Compensation
- Lead-Lag

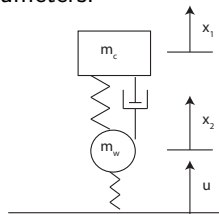
Generalized Root Locus

We may want to know the effect of changing other parameters.

Examples:

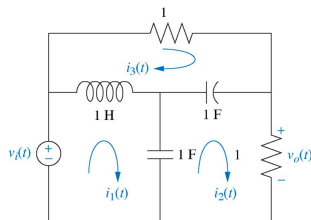
Physics (e.g. Suspension System)

- Spring Constant
- Damper
- Wheel Mass



Circuits (e.g. Toaster, Video Surveillance)

- Resistors
- Capacitors
- Inductors



Maybe there is no control at all!

Root Locus as a General Tool

What do parameters do?

Suspension system: The full TF:

$$\frac{K_2(m_c s^2 + cs + K_1)}{m_c m_w s^4 + c(m_w + m_c)s^3 + (K_1 m_c + K_1 m_w + K_2 m_c)s^2 + cK_2 s + K_1 K_2}$$

The Effect of Damping Constant: c

- No Feedback
- Only examine c
 - ▶ Everything else is 1.

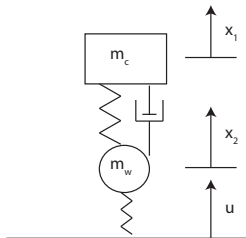
$$G(s) = \frac{s^2 + cs + 1}{s^4 + 2cs^3 + 3s^2 + cs + 1}$$

Where are the Poles?

$$s^4 + 2cs^3 + 3s^2 + cs + 1 = 0$$

$$s^4 + 3s^2 + 1 + c(2s^3 + s)$$

$$= d(s) + cn(s) = 0$$



Looks like the root locus!

Root Locus as a General Tool

What do parameters do?

$$G(s) = \frac{s^2 + cs + 1}{d(s) + cn(s)}$$

- $d(s) = s^4 + 3s^2 + 1$
- $n(s) = 2s^3 + s$

The root locus is the set of roots of

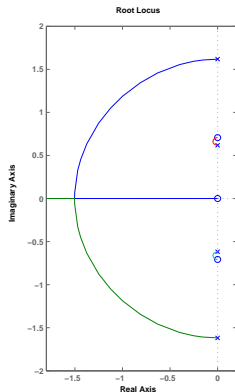
$$d(s) + kn(s)$$

We plot the root locus for

$$G_c(s) = \frac{n(s)}{d(s)} = \frac{2s^3 + s}{s^4 + 3s^2 + 1}$$

Note that G_c is totally fictional!

G_c must still be proper (n is lower degree than d).



The effect of changing c is small.

Root Locus as a General Tool

Suspension Example: Damping Ratio

Root locus tells us:

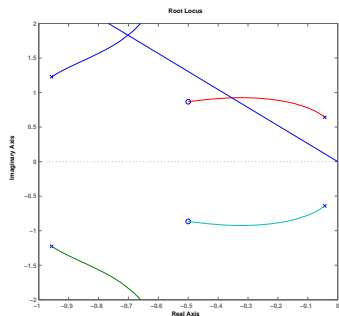
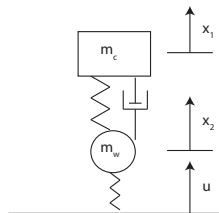
- Changing c won't help with overshoot.
- We need Feedback!

Set $c = 1$ and plot the root locus

$$G(s) = \frac{s^2 + s + 1}{s^4 + 2s^3 + 3s^2 + s + 1}$$

- Examine the gain at
 - ▶ $s_1 = -.3536 + .922i$
 - ▶ $s_2 = -.7 + 1.83i$
- Find Crossover Points
 - ▶ $k = 3.58$
 - ▶ $k = 2.61$

We'll want $k \cong 3$.



Root Locus as a General Tool

Suspension Example: Damping Ratio

Closed Loop Transfer Function:

$$\frac{kG(s)}{1 + kG(s)} = \frac{k(s^2 + cs + 1)}{k(s^2 + cs + 1) + s^4 + 2cs^3 + 3s^2 + cs + 1}$$

Can damping ratio get us to 30% overshoot?

- With feedback
- Set $k = 3$ and plot root locus

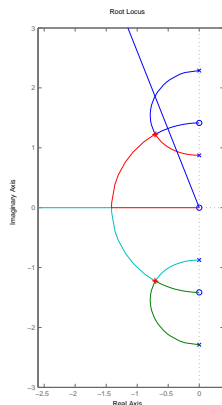
Closed Loop Transfer Function ($k = 3$):

$$G_{kc}(s) = \frac{s^2 + cs + 1}{(s^4 + 6s^2 + 4) + c(2s^3 + 3s + s)}$$

Use `rlocfind` to pick off the best value of c .

Choose the point $s = -.71 + 1.215i$.

- Yields $c = \frac{|d(s)|}{|n(s)|} = 1.414$



Root Locus as a General Tool

Suspension Example: Damping Ratio

Using $c = 1.414$ and $k = 3$, the closed-loop transfer function is

$$\frac{kG_c(s)}{1 + kG_c(s)} = \frac{3s^2 + 4.24s + 3}{s^4 + 2.8s^3 + 6s^2 + 5.7s + 4}$$

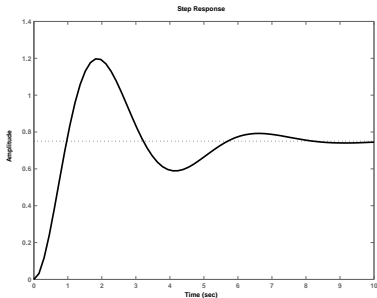
Which has repeated poles at

$$s_{1,2,3,4} = -.71 \pm 1.2i$$

Corresponds to an overshoot of

$$M_p = e^{-\frac{\pi\sigma}{\omega}} = e^{-\frac{.71*\pi}{1.2}} = .18$$

The real overshoot is much bigger!

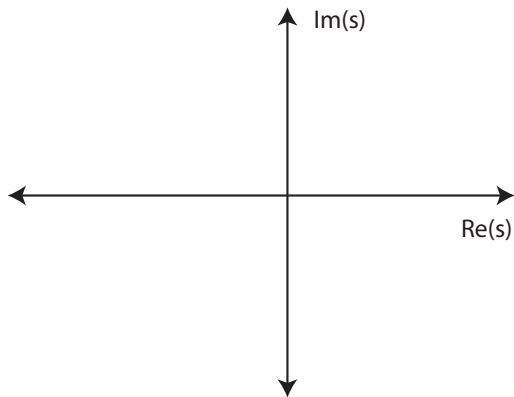


Root Locus as a General Tool

DIY Example

$$G(s) = \frac{s^2 + b^2s + b}{s^3 + (7 + b)s^2 + (12 + b)s + b}$$

Find the optimal value of b .



Limitations of Root Locus

Root Locus isn't perfect

- Can only study one parameter at a time.
- What to do if root locus doesn't go where we want?

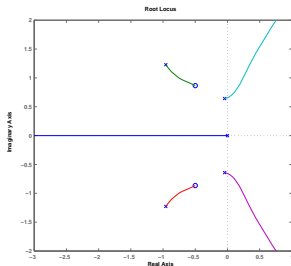
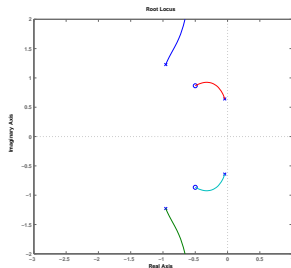
The Suspension Problem:

$$G(s) = \frac{s^2 + s + 1}{s^4 + 2s^3 + 3s^2 + s + 1}$$

Adding a pole at the origin has a negative effect.

Question:

Would adding a zero have a positive effect?



Limitations of Root Locus

The Inverted Pendulum:

$$G(s) = \frac{1}{s^2 + .5}$$

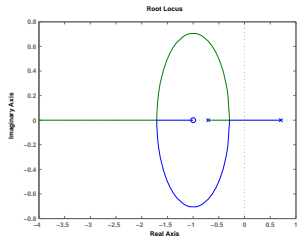
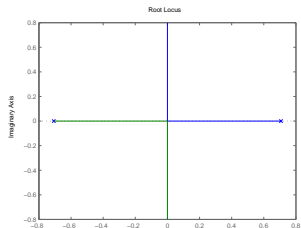
We used PD feedback $K(s) = k(1 + T_D s)$

- Puts a zero at $s = \frac{1}{T_D}$

Conclusion:

- Adding a zero at $z = -1$ improves performance.

Can we generalize this?



Adding Poles and Zeroes

PID control

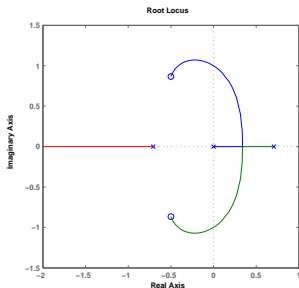
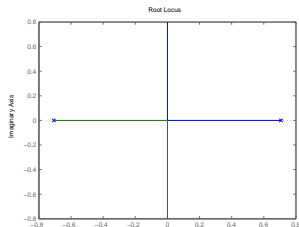
PID feedback:

$$K(s) = k \left(1 + T_i \frac{1}{s} + T_D s \right)$$
$$= k \frac{T_D s^2 + s + T_I}{s}$$

Adds poles and zeros:

- Two zeros: $z_{1,2} = -\frac{1 \pm \sqrt{1 - 4T_D T_I}}{2T_D}$
- One pole: $p = 0$

Question: Is there a systematic way to add poles and zeros?

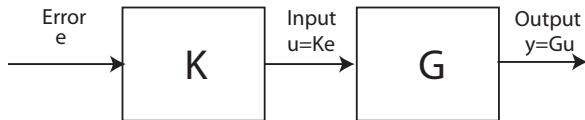


Adding Poles and Zeroes

How?

How To Add a Pole/Zero?

- What does it mean?



Constraint: The plant is fixed.

- $G(s)$ doesn't change.

The pole/zero must come from the controller. e.g.

What is a Controller?

- A system
 - ▶ Maps $e(t)$ to $u(t)$

Adding Poles and Zeroes

Zeros

Goal: Add a Zero

- Like PD control.

Controller:

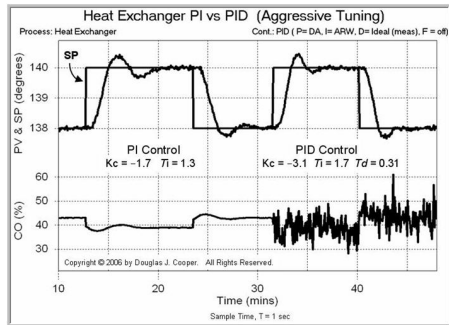
$$K(s) = k(s + z)$$

- **Input to Controller:** $\hat{e}(s)$
- **Output from Controller:**

$$\begin{aligned}\hat{u}(s) &= k(s + z) \\ &= ks\hat{e}(s) + kz\hat{e}(s)\end{aligned}$$

Time-Domain:

$$u(t) = k e'(t) + kz e(t)$$



Problem: Requires differentiation.

$$e'(t) \cong \frac{e(t) - e(t - \tau)}{\tau}$$

Adding a Pole

Goal: Add a pole

Controller:

$$K(s) = k \frac{1}{s + p}$$

Input to Controller: $\hat{e}(s)$

Output from Controller: $\hat{u}(s) = \frac{k}{s+p} \hat{e}(s)$

Internal Variable: x .

- Frequency Domain

$$(s + p)x(s) = ke(s)$$

$$u(s) = x(s)$$

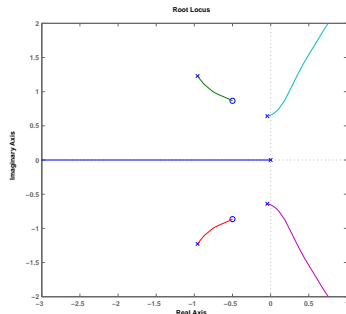
- Time-Domain

$$\dot{x}(t) = -px(t) + ke(t)$$

$$u(t) = x(t)$$

Adding a Pole:

- Requires us to construct a dynamical system whose output is the control.
- Easier than adding a zero, but less useful
 - ▶ Zeros are better for attracting poles away from RHP.



Pole-Zero Compensation

The best way to modify the root locus is by using a pole-zero compensator.

- Adds a zero without differentiation

$$K(s) = k \frac{s - z}{s - p}$$

Input to Controller: $\hat{e}(s)$

Output from Controller: $\hat{u}(s) = k \frac{s-z}{s-p} \hat{e}(s)$

Doing long division:

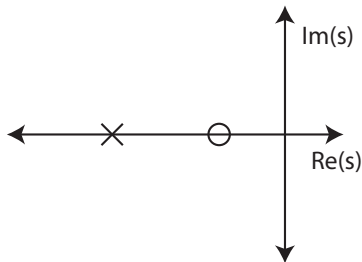
$$\frac{s - z}{s - p} = 1 + \frac{z - p}{s - p}$$

Hence

$$\hat{u}(s) = k \hat{e}(s) + k \frac{z - p}{s - p} \hat{e}(s)$$

Internal Variable:

$$\hat{x}(s) = \frac{k(z - p)}{s - p} \hat{e}(s)$$



Pole-Zero Compensation

Internal Variable:

$$\hat{x}(s) = \frac{k(z - p)}{s - p} \hat{e}(s)$$

- Frequency Domain:

$$(s + p)x(s) = k(z - p)e(s)$$

$$u(s) = x(s) + k\hat{e}(s)$$

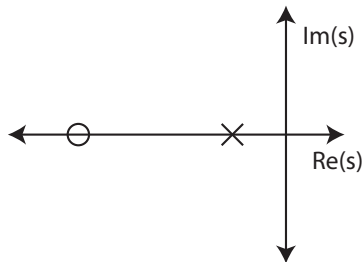
- Time-Domain:

$$\dot{x}(t) = -px(t) + k(z - p)e(t)$$

$$u(t) = x(t) + ke(t)$$

Artificial Dynamics:

- Controller State: $x(t)$
- No differentiation of $e(t)$!
- A zero should always be combined with a pole.



Lead Compensation

Definition 1.

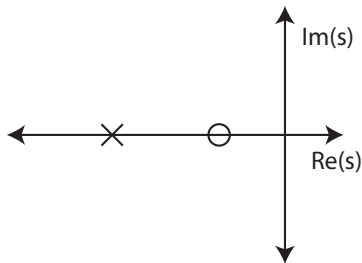
A **Lead Compensator** is a pole-zero compensator

$$K(s) = \frac{s + z}{s + p}$$

where $p < z$.

Used when we really want a zero

- The pole has less effect than the zero.



Lead Compensation

Inverted Pendulum

$$G(s) = \frac{1}{s^2 - .5}$$

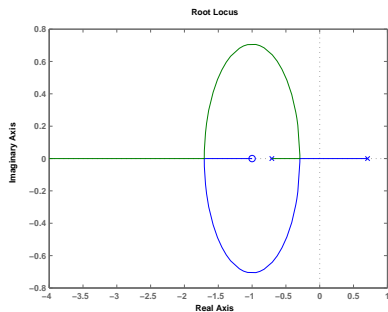


Figure : $K(s) = k(s+1)$

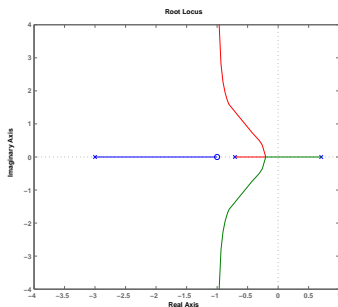


Figure : $K(s) = k \frac{s+1}{s+3}$

Lag Compensation

Definition 2.

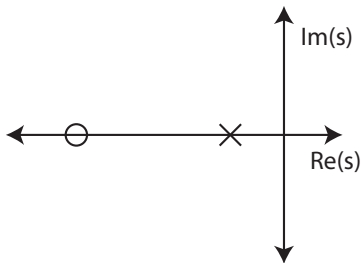
A **Lag Compensator** is a pole-zero compensator

$$K(s) = \frac{s + z}{s + p}$$

where $z < p$.

Used when we really want a pole

- The zero has less effect than the pole.
- Doesn't increase the number of asymptotes.



Lag Compensation

Suspension Problem

$$G(s) = \frac{s^2 + s + 1}{s^4 + 2s^3 + 3s^2 + s + 1}$$

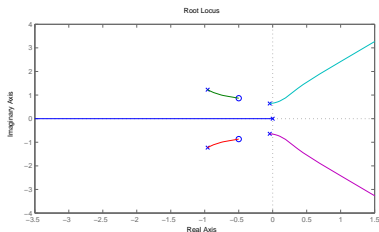


Figure : $K(s) = \frac{k}{s}$

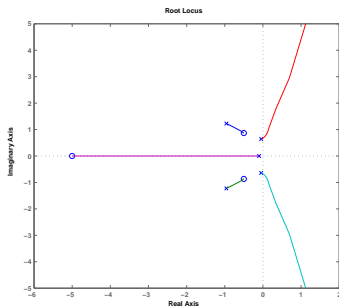


Figure : $K(s) = k \frac{s+5}{s+.1}$

Summary

What have we learned today?

Generalized Root Locus?

- What about changing *OTHER* parameters
- T_D , T_I , et c.
- mass, damping, et c.

Compensation via Root-Locus

- Introduction
- Pole-Zero Compensation
- Lead-Lag

Next Lecture: Pole-Zero Compensation and Notch Filters